

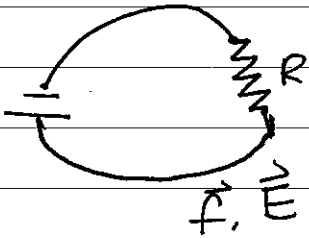
Electrodynamics

No.

Date.

7-1

Electromotive force



As we discussed, charges are pushed through a resistor by \vec{E} .

In steady state, $\vec{v} \times \vec{E} = 0$

\therefore The force by \vec{E} per unit charge

$$\vec{f}_E = \vec{E}$$

and $\oint \vec{f}_E \cdot d\vec{Q} = 0$

There is no net work done by \vec{E} !

Then, what keeps charges moving around the circuit?

Usually, this is provided by some driving sources such as battery.

In this case, charges ... experience:

\vec{f}_S (usually in some portion of circuit)

$$\therefore \vec{f} = \vec{f}_S + \vec{f}_E = \vec{f}_S + \vec{E}$$

is the total driving force.

Note that \vec{f}_s may not be a real force. For instance, for a battery, \vec{f}_s results from chemical energy.

Therefore, what is more important is the quantity

$$\mathcal{E} = \oint \vec{f} \cdot d\vec{e}$$

which is the work done on the charge by all driving sources. \mathcal{E} is called electromotive force, or emf.

$$\therefore \oint \vec{f}_E \cdot d\vec{e} = 0$$

$$\therefore \mathcal{E} = \oint \vec{f}_s \cdot d\vec{e} = \text{work done per unit charge by source}$$

a || b

Inside an ideal source of emf ($f_s \neq 0$)

(from a to b), $f = 0 = \frac{V}{\Delta}$ (i.e. $\Delta = \infty$)
 $\vec{f}_s = -\vec{E}$
 resistanceless battery

$$\therefore \mathcal{E} = \oint \vec{f}_s \cdot d\vec{e} = \int_a^b \vec{f}_s \cdot d\vec{e} = - \int_a^b \vec{E} \cdot d\vec{e} = V_b - V_a$$

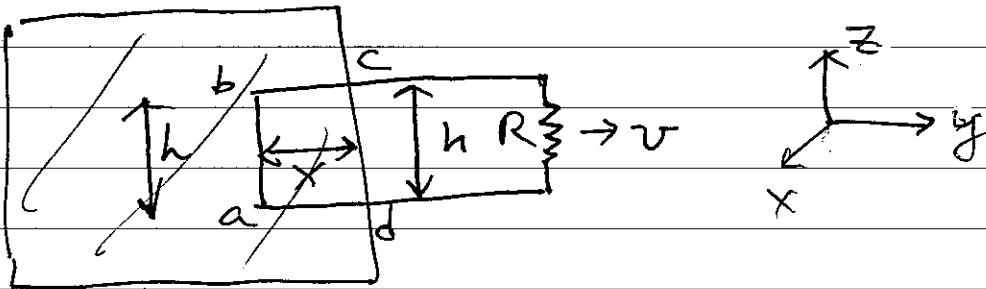
\therefore The battery maintains a voltage difference.

across the battery, that drives currents
flow outside the battery.

Motional emf

In addition to the battery, generators
 are another way to provide emf.

In this case, one exploits emf due
 to motions of conducting wires.



Consider a primitive model for a generator, in
 which a loop with resistance is pulled out
 from a region with uniform magnetic field,
 as shown in the above figure.

If the velocity of the loop is v ,

then charges in segment ab experience a
 magnetic force whose perpendicular
 component is $F_z = qvB$.

Hence

$$f_{\text{mag}}^z = \frac{F_z}{q} = vB$$

$$\mathcal{E} = \oint \vec{f}_{\text{mag}} \cdot d\vec{\ell} = \int_a^b f_{\text{mag}}^z d\ell$$

$$= vBh \quad \text{--- (1)}$$

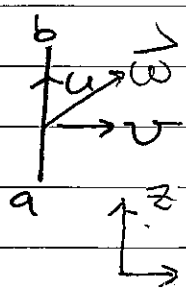
\therefore Lorentz force generates an emf

$\mathcal{E} = vBh$ in this case, It is

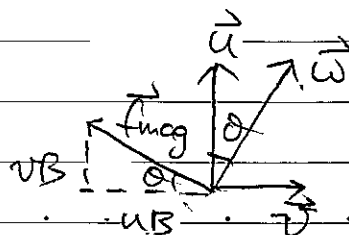
called motional emf.

This seems to stem from the magnetic field but the magnetic force does not do work. Instead, the work is done by the force that pulls the loop!

Similar to the case we mentioned at the beginning of Chapter 5. Charges in ab segment have velocity



$$\vec{v} = v\hat{x} + u\hat{z} \quad \text{--- (2)}$$



$$\begin{aligned} \therefore \vec{f}_{\text{mag}} &= \vec{v} \times (B\hat{y}) \\ &= -vB\hat{x} + Bu\hat{z} \end{aligned}$$

\vec{f}_{mag} still in perpendicular to $\vec{\omega}$

\therefore It does not do work on charges.

However, it yields a force $-uB\hat{x}$.

To counter it, one has to apply a

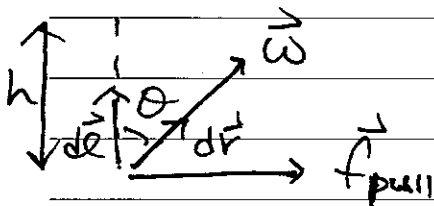
$$\text{force } \vec{f}_{\text{pull}} = uB\hat{x} \quad \dots (3)$$

As a result of \vec{f}_{pull} , the total force ^{acting}

on unit charge is $\vec{f}_{\text{mag}} + \vec{f}_{\text{pull}} = Bv\hat{z}$

does work on charges.

The work done on ^{a unit} charge is



$$W = \int (\vec{f}_{\text{mag}} + \vec{f}_{\text{pull}}) \cdot d\vec{r} \quad \dots (4)$$

$$= \int \vec{f}_{\text{pull}} \cdot d\vec{r}$$

$$\vec{f}_{\text{mag}} \perp d\vec{r}$$

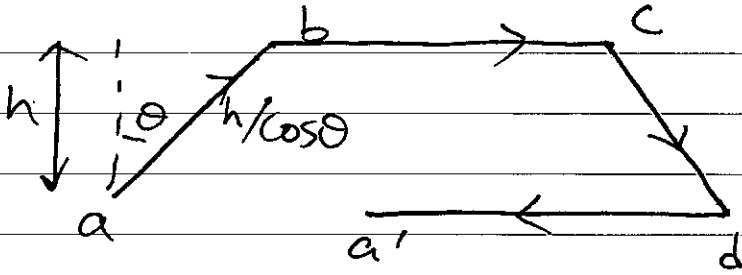
$$= \int f_{\text{pull}} dr \sin\theta$$

$$= f_{\text{pull}} \sin\theta \frac{h}{\cos\theta} = Bh u \tan\theta$$

$$= u \cdot B \cdot h = \Sigma \cdot m \frac{v}{c} \cdot \dots$$

Note that the path of a charge moves

during a period is



$$\therefore \vec{f}_{\text{mag}} + \vec{f}_{\text{pull}} = f_{\text{mag}} \hat{z}$$

$$\therefore (\vec{f}_{\text{mag}} + \vec{f}_{\text{pull}}) \cdot d\vec{r} = f_{\text{mag}} \hat{z} \cdot d\vec{r}$$

$$= f_{\text{mag}} \cdot d\vec{z}$$

$$\therefore W = \oint \vec{f}_{\text{mag}} \cdot d\vec{z}$$

which has the apparent form that \vec{f}_{mag} seems

to do the work but actually it

is done by \vec{f}_{pull} .

Flux rule

It's convenient to note that

$$v = -\frac{dx}{dt} \quad \text{if } x = \text{length that}$$

is inside the magnetic field region

$$\therefore \mathcal{E} = -Bl \frac{dx}{dt} = -\frac{d(Blx)}{dt} \quad \text{--- (6)}$$

Realizing $\oint \vec{B} \cdot d\vec{l} =$ ^{magnetic} flux that the loop

$$\text{encloses} \equiv \Phi \equiv \int_S \vec{B} \cdot d\vec{a}$$

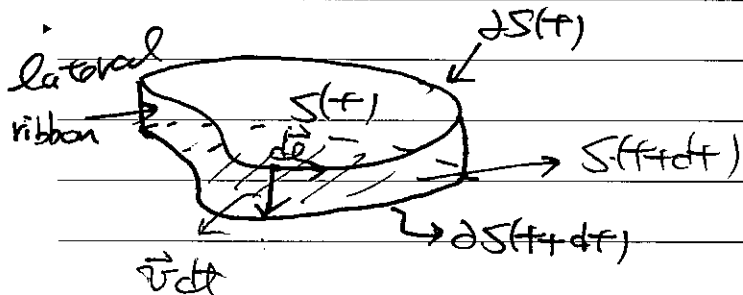
$$\therefore \mathcal{E} = - \frac{d\Phi}{dt} \quad \text{--- (7) which is the flux}$$

rule for the motional emf.

The flux rule can be applied to any general loops which can ^{also} move

in time.

As shown in the



left figure, the

loop at t is

$\partial S(t)$ and it

moves to $\partial S(t+dt)$ at $t+dt$.

Let the lateral area (ribbon) be $\Delta S \equiv S_{\text{ribbon}}$

$$\therefore \vec{\nabla} \cdot \vec{B} = 0 \quad \therefore \int_{S(t+dt)} \vec{B}(t+dt) \cdot d\vec{a}$$

$$= \int_{\text{any surface bounded by } \partial S(t+dt)} \vec{B}(t+dt) \cdot d\vec{a}$$

$$= \int_{S(t)} \vec{B}(t+dt) \cdot d\vec{a} + \int_{S_{\text{ribbon}}} \vec{B}(t+dt) \cdot d\vec{a}$$

For static field, \vec{B} is independent of

t ,

$$\therefore \Phi(t+dt) = \int_{S(t)} \vec{B} \cdot d\vec{a} + \int_{\text{Ribbon}} \vec{B} \cdot d\vec{a} \quad \text{--- (8)}$$

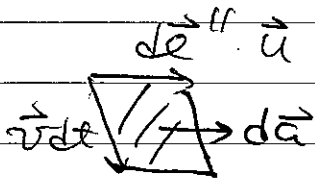
$$= \Phi(t) + \int_{\text{Ribbon}} \vec{B} \cdot d\vec{a}$$

$$\therefore d\Phi = \Phi(t+dt) - \Phi(t)$$

$$= \int_{\text{Ribbon}} \vec{B} \cdot d\vec{a} \quad \text{--- (9)}$$

Now for each point on $S(t)$, it moves

with some velocity \vec{v} to $S(t+dt)$



$$\therefore d\vec{a} = \vec{v} dt \times d\vec{e}$$

L. - (10)

\therefore Eqs (9) & (10) imply

$$\frac{d\Phi}{dt} = \oint \vec{B} \cdot (\vec{v} \times d\vec{e}) \quad \text{--- (11)}$$

If the tangential velocity of charge is \vec{u} ,

$\vec{u} \parallel d\vec{e}$, the total velocity of the charge

$$\vec{w} = \vec{u} + \vec{v} \quad \therefore \vec{v} \times d\vec{e} = \vec{w} \times d\vec{e}$$

($\vec{v} \times d\vec{e} = 0$)

Hence

$$\frac{d\Phi}{dt} = \oint \vec{B} \cdot \vec{\omega} \times d\vec{e}$$

$$= \oint \vec{B} \times \vec{\omega} \cdot d\vec{e}$$

$$= - \oint \underbrace{\vec{\omega} \times \vec{B}} \cdot d\vec{e}$$

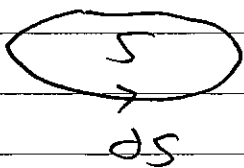
magnetic force per unit charge

$$\therefore \frac{d\Phi}{dt} = - \oint \vec{f}_{\text{mag}} \cdot d\vec{e} = -\Sigma$$

$$\therefore \Sigma = - \frac{d\Phi}{dt} \text{ is generally correct } \text{---} \textcircled{12}$$

Direction: $\Phi = \int_S \vec{B} \cdot d\vec{a}$ is defined to

be positive if $\vec{B} \parallel d\vec{a}$.



\therefore Right-hand rule:

thumb $\hat{=} d\vec{a}$

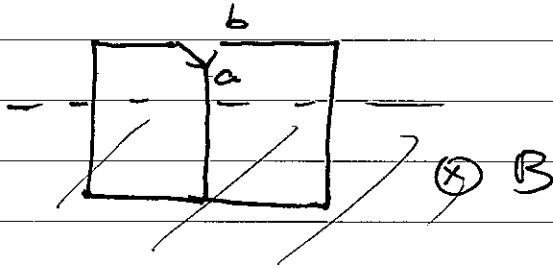
4-fingers = orientation of $d\vec{s} \oint_{\partial S}$

Note that the motional emf must involve

the motion of segments in B field. Switches

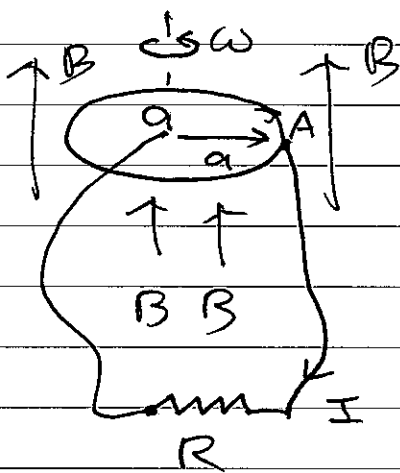
or sliding contacts are not included.

A false example :



Switch a to b does n't yield any emf!

Example Farady disk



A metal disk of radius a rotates with angular velocity ω in a uniform B field.

Find the current I in R .

Solution: This problem can not be solved by flux rule as there is no definite conducting loop.

For any point between OA , $0 < p < a$,

$v = \omega p$, the magnetic force per unit

charge is $\vec{f}_{\text{mag}} = \vec{v} \times \vec{B} = \omega p B \hat{p}$

$$\therefore \mathcal{E} = \int_0^a f_{\text{mag}} dp = \omega B \int_0^a p dp = \frac{1}{2} \omega B a^2$$

$$\therefore I = \frac{\mathcal{E}}{R} = \frac{\omega B a^2}{2R}$$

Eddy current

In the case when there is no clear conducting loop; free charge carriers will

form their loops by themselves. This

is the case when one moves a piece

of metal (say aluminum) towards or

away from a magnet.

In this case, there are eddy currents

in which charges in the metal move

in response to emf generated by motion.

(or changing of magnetic fields, which will be discussed later.) The motion of charges

forms swirling current, called eddy currents. The eddy current will impose

damping to the motion of metal. To

avoid generation of eddy currents, one way

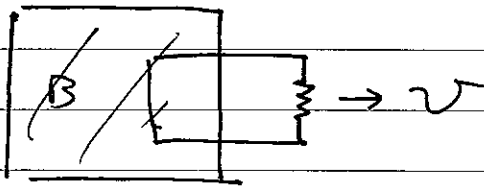
is to cut the path of current by introducing slots in the metal disk.

Electromagnetic induction

Faraday's law

In 1831, Faraday performed a series of experiments that contains 3 most important setup of experiments that lead to the induction of emf.

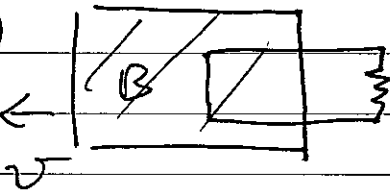
(i)



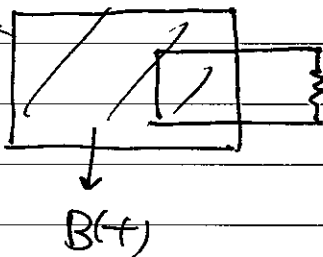
motional emf

$$\mathcal{E} = - \frac{d\Phi}{dt}$$

(ii)



(iii)



(ii) is relative motion version of (i)

\therefore the same result $\mathcal{E} = - \frac{d\Phi}{dt}$ is obtained

In (iii), one also obtains

$$\mathcal{E} = - \frac{d\Phi}{dt}$$

For (ii) & (iii), there is no motion of charges.

Therefore, it's not due to the magnetism.

force $\propto \vec{v} \times \vec{B}$!

What is responsible for \mathcal{E} ?

Faraday :

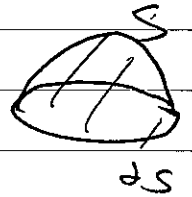
A changing magnetic field induces
an electric field --- (13)

The induced electric field \vec{E} obeys

$$\mathcal{E} = \oint_{\partial S} \vec{E} \cdot d\vec{e} = - \frac{d\Phi}{dt} \quad \text{--- (14)}$$

Since $\Phi = \int_S \vec{B} \cdot d\vec{a}$, $\frac{d\Phi}{dt} = \int \frac{\partial \vec{B} \cdot \vec{a}}{\partial t} d\vec{a}$

$$\therefore \oint_{\partial S} \vec{E} \cdot d\vec{e} = - \int_S \frac{\partial \vec{B}}{\partial t} \cdot d\vec{a} \quad \text{--- (15)}$$



This is Faraday's law in the
integral form. Note that S & ∂S are
fixed and don't depend on time!

Using the Stokes's theorem,

$$\oint_{\partial S} \vec{E} \cdot d\vec{e} = \int_S \vec{\nabla} \times \vec{E} \cdot d\vec{a}$$

taking $S \rightarrow 0$, one gets (in contrast to $\vec{\nabla} \times \vec{E} = 0$
in electrostatics)

$$\vec{\nabla} \times \vec{E} = - \frac{\partial \vec{B}}{\partial t} \quad \text{--- (16)}$$

Which is the Faraday's law in differential
form.

Note that many people called

$$\mathcal{E} = -\frac{d\Phi}{dt} \text{ as the}$$

Faraday's law. However, $\mathcal{E} = -\frac{d\Phi}{dt}$

includes (i)-(iii) cases while ^{only} eqs (15) & (16)

accounts for (ii) (iii)

Case (i) has different mechanism and

it is still included in $\mathcal{E} = -\frac{d\Phi}{dt}$.

Therefore, eqs. (15) or (16) deserve the name of

Faraday's law.

The coincidence of the ^{same} form

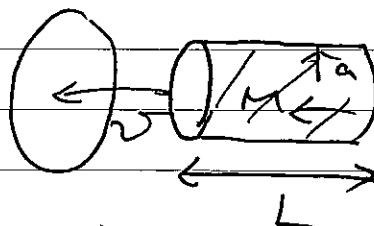
$\mathcal{E} = -\frac{d\Phi}{dt}$ for (i) & (ii)-(iii) has deeper

reason related to special relativity.

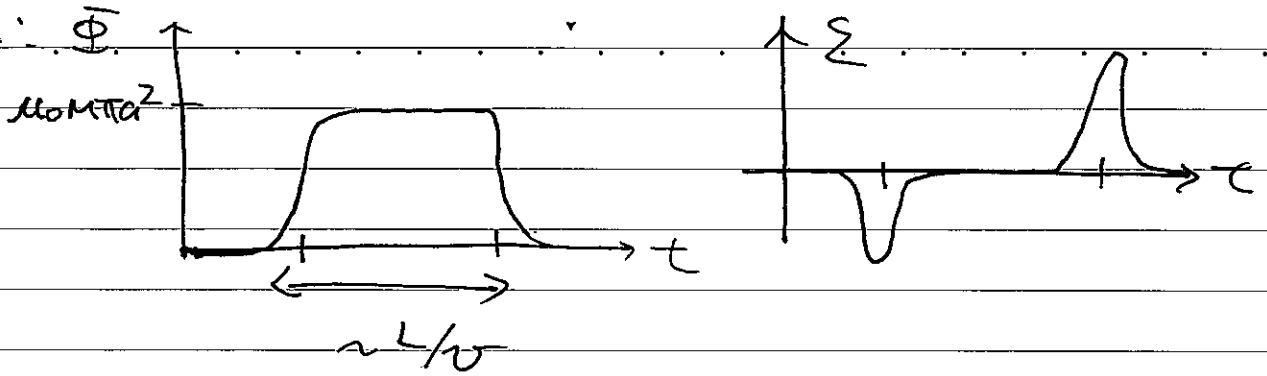
We will come back to it later in

Chapter 12.

Example



$$\vec{B} = \mu_0 \vec{M}, \text{ close to ring: } \Phi_B \approx \pi a^2 \cdot \mu_0 M$$



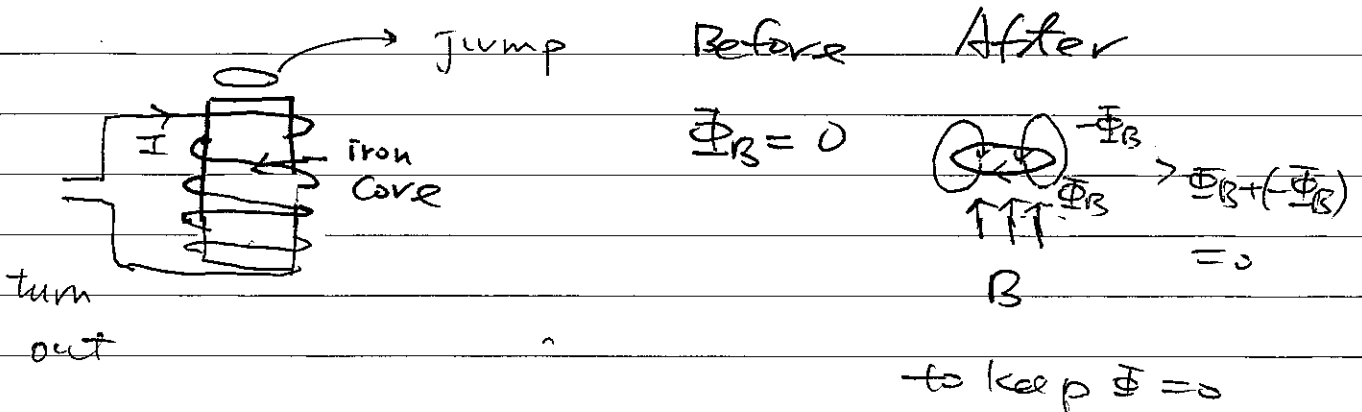
Lenz's law = direction of current flow

In the Faraday's law, it is not easy to keep track of the signs and induced current.

Lenz discovered an easy rule to get the right direction. That is,

the induced current tries to oppose the change of the flux.

Example: Jumping ring



The induced Electric field

* In terms of vector potential

Faraday's law generalizes the

$$\text{electrostatic } \vec{\nabla} \times \vec{E} = 0$$

$$\text{to } \vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t} \quad \dots (17)$$

$$\because \vec{B} \text{ still satisfies } \vec{\nabla} \cdot \vec{B} = 0$$

$\therefore \vec{B} = \vec{\nabla} \times \vec{A}$ is still correct even if

\vec{B} depends on time. i.e. $\vec{B}(\vec{r}, t) = \vec{\nabla} \times \vec{A}(\vec{r}, t)$

$$\therefore \vec{\nabla} \times \vec{E} = -\frac{\partial}{\partial t} \vec{\nabla} \times \vec{A} = \vec{\nabla} \times \left(-\frac{\partial \vec{A}}{\partial t} \right)$$

Therefore, the generalization becomes

$$\vec{\nabla} \times \left(\vec{E} + \frac{\partial \vec{A}}{\partial t} \right) = 0$$

Hence $\vec{E} + \frac{\partial \vec{A}}{\partial t} = \nabla(\dots) \equiv -\nabla V$, $V = \text{scalar potential}$

$$\therefore \vec{E} = -\nabla V - \frac{\partial \vec{A}}{\partial t} \quad \dots (18)$$

which generalizes the relation $\vec{E} = -\nabla V$.

$$\text{Since } \oint_{\partial S} \vec{E} \cdot d\vec{s} = \int_S \vec{\nabla} \times \left(-\nabla V \right) \cdot d\vec{a} = 0,$$

$$\oint_{\partial S} \left(-\nabla V \right) \cdot d\vec{s} = 0$$

$$\oint_S \vec{E} \cdot d\vec{e} = \oint_S \left(-\nabla V - \frac{d\vec{A}}{dt} \right) \cdot d\vec{e}$$

$$= \oint_S -\frac{d\vec{A}}{dt} \cdot d\vec{e}$$

$$= -\frac{d}{dt} \oint_S \vec{A} \cdot d\vec{e} = -\frac{d}{dt} \Phi$$

Hence one sees that it is the participation of \vec{A} in \vec{E} so that

$$\oint_S \vec{E} \cdot d\vec{e} \neq 0 \quad \text{On one hand, } \nabla \times \vec{A} \dots$$

produces \vec{B} . On the other hand, $-\frac{d\vec{A}}{dt}$

gives rise to the induced electric field.

One can ^{also} compare the case $\rho = 0, \vec{J} \neq 0$

$$\vec{\nabla} \cdot \vec{E} = 0 \quad \nabla \times \vec{E} = -\frac{d\vec{B}}{dt}$$

$$\vec{\nabla} \cdot \vec{B} = 0 \quad \nabla \times \vec{B} = \mu_0 \vec{J}$$

$-\frac{d\vec{B}}{dt}$ plays a similar role to \vec{J} .

$$\therefore \vec{E} = \frac{1}{4\pi} \int \frac{-\frac{d\vec{B}(r')}{dt} \times \vec{r}}{r^2} dz'$$

$$= -\frac{d}{dt} \frac{1}{4\pi} \int \frac{\vec{B} \times \vec{r}}{r^2} dz'$$

$$\therefore \oint \vec{E} \cdot d\vec{e} = -\frac{d}{dt} \oint dF \cdot \int \frac{\vec{B}(\vec{r}, t) \times (\vec{r} - \vec{r}')}{4\pi |\vec{r} - \vec{r}'|^3} dz'$$

L. (19)

Now, in the case of magnetostatics,

$$\vec{B}(\vec{r}) = \frac{\mu_0}{4\pi} \int \frac{\vec{J}(\vec{r}') \times (\vec{r} - \vec{r}')}{|\vec{r} - \vec{r}'|^3} dz' \quad (\text{Biot-Savart})$$

$$\oint \vec{B} d\vec{r} = \mu_0 I = \mu_0 \int \vec{J} \cdot d\vec{a} \quad \text{--- (20)}$$

$$\therefore \oint d\vec{r} \cdot \int \frac{\vec{B}(\vec{r}, t) \times (\vec{r} - \vec{r}')}{4\pi |\vec{r} - \vec{r}'|^3} dz' = \int \vec{B} \cdot d\vec{a} = \Phi$$

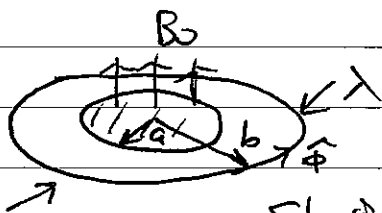
(20) implies

$$\therefore \text{Eq. (19) implies } \oint \vec{E} \cdot d\vec{e} = -\frac{d\Phi}{dt}$$

\(\therefore\) we see $-\frac{d\vec{B}}{dt}$ plays a role similar to \vec{J} .

Example

B_0 is turned off



suspended

Solution:

How ^{does} the ring rotate?

Final angular momentum?

Let z 's law: $\rightarrow E$

Symmetry implies $\vec{E} = E \hat{\phi}$

$$\therefore \oint \vec{E} \cdot d\vec{e} = 2\pi b E = -\frac{d\Phi}{dt} = -\pi a^2 \frac{dB}{dt}$$

$$\therefore \vec{E} = -\frac{a^2}{2b} \frac{dB}{dt} \hat{\phi}$$

For a segment $d\vec{r}$, it experiences

$$\text{a torque } d\vec{z} = \vec{r} \times \underbrace{[\lambda d\vec{r} \vec{E}]}_{d\vec{F}}$$

$$\therefore dz = \lambda b E de = -\lambda b \frac{q^2}{2b} \frac{dB}{dt} de$$

\therefore total torque

$$= -\lambda b \frac{q^2}{2b} \frac{dB}{dt} \oint de = -\lambda b \pi a^2 \frac{dB}{dt}$$

$$\therefore \text{torque} = \frac{dL}{dt}$$

$$\therefore L = \int -\lambda b \pi a^2 \frac{dB}{dt} dt = \lambda \pi a^2 \pi B_0$$

The ring gets angular momentum!

where does it come from?

(We shall discuss it later.)

Quasi-static

In principle, when $\frac{d\vec{r}}{dt} \neq 0$, \vec{B} is time

dependent. In this case, if $\vec{\nabla} \times \vec{B} = \mu_0 \vec{J}$

is still correct. \vec{J} must be time-dependent.

Hence $\frac{d\vec{J}}{dt} \neq 0$. We are no longer

in the steady state!

However, if the change is ^{not} extremely rapid, the relation $\vec{\nabla} \times \vec{B} = \mu_0 \vec{J}$ (Ampere's law) & Biot-Savart law are approximately correct with negligible errors.

In this case, we can still apply

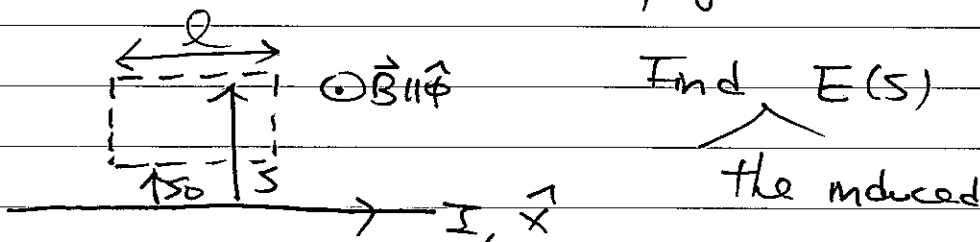
the Ampere's law & Biot-Savart law.

The regime where magnetostatics is

applicable while $\vec{\nabla} \times \vec{E} = -\frac{d\vec{B}}{dt}$ & $\vec{\nabla} \cdot \vec{E} = \rho/\epsilon_0$

are applicable is called quasi-static.

Example. Slow varying current $I(t)$



Solution: quasi-static. $\vec{B} = \frac{\mu_0 I}{2\pi r} \hat{\phi}$

$\vec{E} \parallel \hat{x}$. Consider a loop shown in the above

$$\oint \vec{E} \cdot d\vec{e} = E(s) \cdot 2\pi s - E(s) \cdot 2\pi s = - \frac{d}{dt} \int \vec{B} \cdot d\vec{a}$$

$$= - \frac{\mu_0}{2\pi} \frac{dI}{dt} \int_s^s \frac{1}{s'} ds' \cdot 2\pi$$

$$= - \frac{\mu_0}{2\pi} \frac{dI}{dt} (\ln s - \ln s_0)$$

$$\therefore \vec{E}(s) = \left(\frac{\mu_0}{2\pi} \frac{dI}{dt} \ln s + \underbrace{\frac{\mu_0}{2\pi} \frac{dI}{dt} \ln s_0}_{E(s_0)} \right) \hat{z}$$

K

K is an s-independent constant but K

depends on t.

K depends on the whole history of $I(t)$

as it is reflected that $E(s)$ is not known. That is, to determine $E(s)$, one has to know how $I(t)$ is turned on and changes from the beginning. We shall discuss it later in Chapter 10.

Note that it may seem that $E \rightarrow \infty$ as

$s \rightarrow \infty$. However, as we shall see later,

E field can not be propagated faster than

the speed of light. Hence, the change of I

Can not propagate faster than c .

For a given period of Δt , after the change of I , for

$s > \Delta t c$, the point at s does not know that I changes.

Hence, the quasi-static ΓS

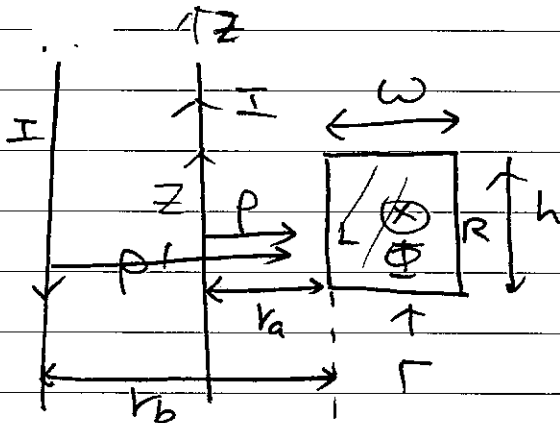
correct (i.e. ^{point} s knows the change of I)

only when $s \ll c\tau$. --- (21)

where τ is the characteristic time scale

that I changes substantially!

Example I : slow varying, $\frac{dI}{dt}$



Find emf in the loop Γ

$$\mathcal{E} = \oint \vec{E} \cdot d\vec{e}$$

$$= \frac{-\lambda}{2\pi\epsilon_0} \ln r + \text{const.}$$

Solution: $\vec{A} = \frac{\mu_0}{4\pi} \int \frac{I d\vec{z}}{|\vec{r}-\vec{z}'|} \Leftrightarrow V = \frac{1}{4\pi\epsilon_0} \int \frac{\lambda dz'}{|\vec{r}-\vec{z}'|}$

$$\therefore \vec{A} = A \hat{z}, \quad A = \frac{-\mu_0 I}{2\pi} \ln r + \text{const for}$$

each current line.

$$\therefore A = \frac{\mu_0 I}{2\pi} \ln \frac{r}{a}$$

\therefore On the L side of loop Γ ,

$$\vec{A}_L = \frac{\mu_0 I}{2\pi} \ln \frac{r_b}{r_a} \hat{z}$$

on the R side of loop Γ ,

$$\vec{A}_R = \frac{\mu_0 I}{2\pi} \ln \frac{r_b + w}{r_a + w} \hat{z}$$

Using $\vec{E} = -\frac{d\vec{A}}{dt}$, one gets

$$\vec{E}_L = -\frac{\mu_0}{2\pi} \frac{dI}{dt} \ln \frac{r_b}{r_a} \hat{z}$$

$$\vec{E}_R = -\frac{\mu_0}{2\pi} \frac{dI}{dt} \ln \left(\frac{r_b + w}{r_a + w} \right) \hat{z}$$

$$\oint \vec{E} \cdot d\vec{e} = -\frac{\mu_0}{2\pi} \frac{dI}{dt} \ln \frac{r_b}{r_a} h + \frac{\mu_0}{2\pi} \frac{dI}{dt} \ln \frac{r_b + w}{r_a + w} h$$

$$= -\frac{\mu_0 h}{2\pi} \frac{dI}{dt} \ln \left[\frac{r_b}{r_a} \frac{r_a + w}{r_b + w} \right]$$

$$= -\frac{d\Phi}{dt}$$

$$\Phi = \frac{\mu_0 I}{2\pi} \left(\int_{r_a}^{r_a + w} \frac{h dp}{p} - \int_{r_b}^{r_b + w} \frac{h dp}{p} \right)$$

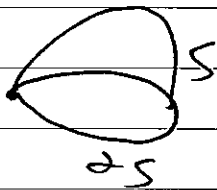
$$= \frac{\mu_0 h I}{2\pi} \ln \frac{r_b (r_a + w)}{r_a (r_b + w)}$$

Induce electric field in a moving system.

The Faraday law (eq. (14)) is valid

when the loop ∂S is fixed, and it's

not moving:

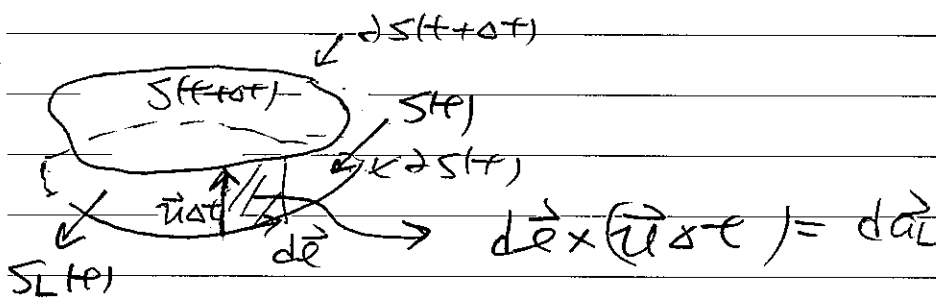


$$\oint_{\partial S} \vec{E} \cdot d\vec{e} = - \int_S \frac{d\vec{B}}{dt} \cdot d\vec{a}$$

L - (15)

When the loop ∂S is moving, eq. (14) has

to be modified.



$$\text{First, } \frac{d\Phi}{dt} = \frac{1}{dt} \left[\int_{S(t+dt)} \vec{B}(t+dt) \cdot d\vec{a} - \int_{S(t)} \vec{B}(t) \cdot d\vec{a} \right]$$

L - (22)

$\therefore S(t+dt) + S(t) + S_L(t)$ forms a close surface.

$$\int_{S(t+dt)} \vec{B}(t+dt) \cdot d\vec{a} - \int_{S(t)} \vec{B}(t+dt) \cdot d\vec{a} + \int_{S_L(t)} \vec{B}(t+dt) \cdot d\vec{a}$$

$$= \int_V \vec{\nabla} \cdot \vec{B}(t+dt) dV = 0$$

$$\therefore \int_{S(t+\Delta t)} \vec{B}(t+\Delta t) \cdot d\vec{a} - \int_{S(t)} \vec{B}(t+\Delta t) \cdot d\vec{a} + \Delta t \int_{\partial S(t)} \vec{B}(t+\Delta t) \cdot d\vec{e} \times \vec{u} = 0$$

L... (22)

$$\text{Now, } \int_{\partial S(t)} \vec{B}(t+\Delta t) \cdot d\vec{e} \times \vec{u} = \int_{\partial S(t)} \vec{u} \times \vec{B}(t+\Delta t) \cdot d\vec{e}$$

$$= \int_{S(t)} \vec{\nabla} \times (\vec{u} \times \vec{B}(t+\Delta t)) \cdot d\vec{a}$$

$$= \int_{S(t)} \vec{\nabla} \times (\vec{u} \times \vec{B}(t)) \cdot d\vec{a} + O(\Delta t)$$

$$\vec{B}(t+\Delta t) = \vec{B}(t) + \Delta t \frac{d\vec{B}(t)}{dt}$$

Eq. (22) becomes

$$\int_{S(t+\Delta t)} \vec{B}(t+\Delta t) \cdot d\vec{a} - \int_{S(t)} \vec{B}(t) \cdot d\vec{a}$$

$$= \Delta t \left[\int_{S(t)} \frac{d\vec{B}(t)}{dt} \cdot d\vec{a} - \int_{S(t)} \vec{\nabla} \times (\vec{u} \times \vec{B}(t)) \cdot d\vec{a} \right]$$

\(\therefore\) Eqs. (22) & (23) imply

$$\frac{d\Phi}{dt} = \int_S \frac{d\vec{B}}{dt} \cdot d\vec{a} - \int_S \vec{\nabla} \times (\vec{u} \times \vec{B}(t)) \cdot d\vec{a}$$

L... (24)

Now, \(\Phi\) is the flux that goes with the loop

Faraday's law implies that

$$-\frac{d\Phi}{dt} = \oint \vec{E} \cdot d\vec{e} \quad \vec{E} \text{ measured in the}$$

frame that moves dS with \vec{u} . --- (25)

$$\text{Eq. (25)} \rightarrow \frac{d\Phi}{dt} = \int_S \vec{\nabla} \times \vec{E} \cdot d\vec{a}$$

\therefore Combined with eq. (20), one gets

$$\int_S \vec{\nabla} \times \vec{E} \cdot d\vec{a} = \int_S \left[\vec{\nabla} \times (\vec{v} \times \vec{B}) - \frac{d\vec{B}}{dt} \right] \cdot d\vec{a}$$

taking $S \rightarrow 0$, one gets

$$\vec{\nabla} \times \vec{E} = -\frac{d\vec{B}}{dt} + \vec{\nabla} \times (\vec{v} \times \vec{B}) \quad \dots (26)$$

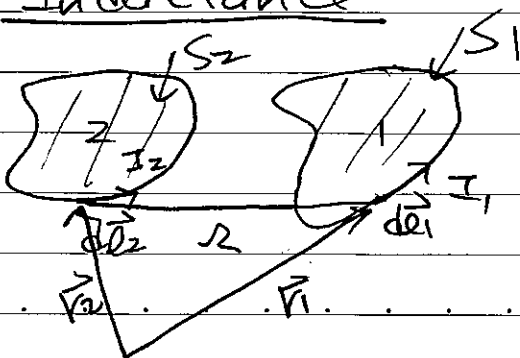
Where \vec{E} is measured in the frame that moves with \vec{v} relative to the frame where \vec{B} is measured.

Example. moving conductor:

\vec{B} measured in lab frame

\vec{E} measured in conductor frame

Inductance

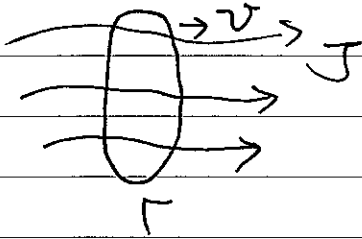


Induced electric field can be induced by other current loops.

Consider two current loops, 1 & 2.

Example: Alfvén's theorem

In a perfect conducting fluid, the magnetic flux through any closed loop moving with the fluid is constant.



Solution: $\therefore \vec{J}_f = \sigma(\vec{E} + \vec{v} \times \vec{B})$

For perfect conducting fluids, $\sigma \rightarrow \infty$

$\therefore \vec{E} + \vec{v} \times \vec{B} = 0, \vec{E} = -\vec{v} \times \vec{B}$

$\therefore \nabla \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$ becomes

$\frac{\partial \vec{B}}{\partial t} - \nabla \times (\vec{v} \times \vec{B}) = 0$

Eq. (24) \vec{u} replaced by \vec{v}

$\therefore \frac{d\Phi}{dt} = 0 \quad \therefore \Phi = \text{constant}$

There is no ^{induced} electric field if one moves with the ring. As we shall see later, in a moving frame, \vec{B} will generate an electric field $\vec{v} \times \vec{B}$ that cancels $\vec{E} = -\vec{v} \times \vec{B}$.

The flux of loop 2 due to current in loop 1

is

$$\Phi_{21} = \int_{S_2} \vec{B}_2 \cdot d\vec{a}_2$$

$\vec{B}_2 = \vec{\nabla} \times \vec{A}_2$ is the magnetic field due to loop 1,

$$\vec{A}_2 = \frac{\mu_0 I_1}{4\pi} \oint_I \frac{d\vec{r}_1}{r_2}$$

$$\therefore \Phi_{21} = \int_{S_2} \vec{\nabla} \times \vec{A}_2 \cdot d\vec{a}_2$$

$$= \oint_2 \vec{A}_2 \cdot d\vec{r}_2$$

$$= \frac{\mu_0 I_1}{4\pi} \oint_I \frac{d\vec{r}_1}{r_2} \cdot \oint_2 d\vec{r}_2$$

$$= \left(\frac{\mu_0}{4\pi} \oint_I \oint_2 \frac{d\vec{r}_1 \cdot d\vec{r}_2}{|\vec{r}_1 - \vec{r}_2|} \right) I_1 \equiv M_{21} I_1$$

Similarly, $\Phi_{12} = \left(\frac{\mu_0}{4\pi} \oint_2 \oint_1 \frac{d\vec{r}_2 \cdot d\vec{r}_1}{|\vec{r}_1 - \vec{r}_2|} \right) I_2 \equiv M_{12} I_2$

Clearly, $M_{12} = M_{21} = \frac{\mu_0}{4\pi} \oint_1 \oint_2 \frac{d\vec{r}_1 \cdot d\vec{r}_2}{|\vec{r}_1 - \vec{r}_2|} \equiv M \quad (27)$

= mutual inductance between two circuit

Eq. (27) is the Neumann equation.

Under quasi-static condition, the change of

I_1 will induce emf in loop 2.

$$\mathcal{E}_2 = - \frac{d\Phi_{21}}{dt} = -M \frac{dI_1}{dt}$$

$$\mathcal{E}_1 = - \frac{d\Phi_{12}}{dt} = -M \frac{dI_2}{dt}$$

\therefore M characterizes emf induced between loops 1 & 2.

Self-inductance

Similarly, change of current in one loop also induces change of flux in the

same loop. \therefore One expects $\Phi = LI \dots (2P)$

$$\mathcal{E} = - \frac{d\Phi}{dt} = -L \frac{dI}{dt} \dots (2P)$$

L is called self inductance.

For MKS system, $\mathcal{E} = \text{Volt}$, $dI = \text{Ampere}$

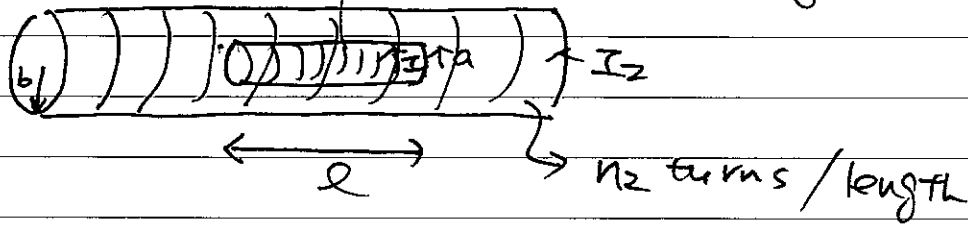
$$\therefore [L] = \text{Volt} \cdot \text{sec} / \text{Ampere} = \text{Henry}$$

Both M & L are intrinsically positive and

are characterized only by geometry of circuits.

Example: Mutual inductance between two

axial solenoids
of turns/length



Find the mutual inductance

Solution: $B_2 = \mu_0 n_2 I_2$

$$\Phi_{12} = B_2 \cdot \pi a^2 \times n_1 l$$

$$= \mu_0 \pi a^2 n_1 n_2 l I_2$$

$$\therefore M_{12} = \mu_0 \pi a^2 n_1 n_2 l$$

$$B_1 = \mu_0 n_1 I_1$$

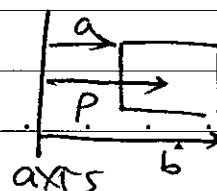
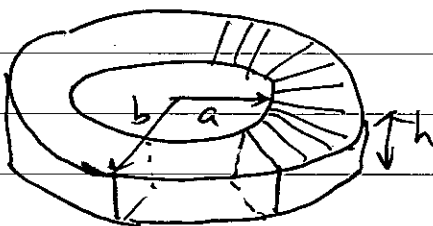
$$\Phi_{21} = B_1 \cdot \underline{\pi a^2} \cdot n_2 l$$

$$= \mu_0 \pi a^2 n_1 n_2 l I_1$$

$$\therefore M_{21} = \mu_0 \pi a^2 n_1 n_2 l = M_{12}$$

Example Self-inductance of a toroidal coil

From the previous example



$$B = \frac{\mu_0 N I}{2\pi r} \quad N = \text{total \# of coil}$$

$$\begin{aligned} \therefore \Phi &= N \times \int \vec{B} \cdot d\vec{a} = \frac{\mu_0 N^2 I h}{2\pi} \int_a^b \frac{dr}{r} \\ &= \frac{\mu_0 N^2 I h}{2\pi} \ln \frac{b}{a} = LI \end{aligned}$$

$$L = \frac{\mu_0 N^2 h}{2\pi} \ln \frac{b}{a}$$

Energy in magnetic fields

It takes a certain amount of energy to set a current in a circuit.

○ We shall assume that the circuit can be described by Ohmic law,

\therefore The current is \vec{J}_f and obeys

$$\vec{J}_f = \sigma \vec{E}$$

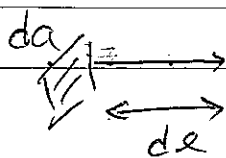
$$\text{With } \vec{E} = -\nabla V - \frac{d\vec{A}}{dt} \quad \dots (30)$$

Now, $-\frac{d\vec{A}}{dt}$ is the induced field. To

maintain the current, the source (battery)

maintains a difference in V .

\vec{v} (distance moved in a unit time)
 \therefore Across $d\vec{e}$, the work done per unit time



$$dW = \dots = \nabla V \cdot d\vec{e} \underbrace{J_f da}_{I_f}$$

$$\therefore \int \vec{J} \cdot d\vec{e} = \int \vec{J} \cdot d\vec{e}$$

$$da d\vec{e} = dz$$

\therefore total work per unit time done

$$\frac{dW}{dt} = \int -\vec{\nabla} \cdot \vec{J} \cdot d\vec{e}$$

$$= \int \left(\vec{E} + \frac{d\vec{A}}{dt} \right) \cdot \vec{J} \cdot d\vec{e}$$

eg. (30)

$$= \int \frac{J^2}{\sigma} dz + \int \frac{d\vec{A}}{dt} \cdot \vec{J} \cdot d\vec{e}$$

--- (31)

The first term is the power or Joule heat.
↑
spent in ohmic

The 2nd term is the work done by the source against the induced emf, and generates the magnetic field.

\therefore we shall see the work for magnetic field

W_m :

$$\frac{dW_m}{dt} = \int \frac{d\vec{A}}{dt} \cdot \vec{J} \cdot d\vec{e} \dots (32)$$

In the absence of magnetic materials,

$$\vec{\nabla} \times \vec{B} = \mu_0 \vec{J}_f$$

$$\therefore \frac{dW_m}{dt} = \frac{1}{\mu_0} \int \frac{\partial \vec{A}}{\partial t} \cdot \vec{\nabla} \times \vec{B} \, d\tau$$

Using $\vec{\nabla} \cdot (\vec{A} \times \vec{B}) = \vec{B} \cdot \vec{\nabla} \times \vec{A} - \vec{A} \cdot \vec{\nabla} \times \vec{B}$,

One gets

$$\frac{dW_m}{dt} = \frac{1}{\mu_0} \int \left[\vec{B} \cdot \nabla \times \frac{\partial \vec{A}}{\partial t} - \vec{\nabla} \cdot \left(\frac{\partial \vec{A}}{\partial t} \times \vec{B} \right) \right] d\tau$$

$$= \frac{1}{\mu_0} \int_V \vec{B} \cdot \frac{\partial \vec{B}}{\partial t} \, d\tau + \oint_S \frac{1}{\mu_0} \vec{B} \times \frac{\partial \vec{A}}{\partial t} \cdot d\vec{a}$$

When $S \rightarrow \infty$ ($r \rightarrow \infty$), $\therefore B \sim \frac{1}{r^3}$, $\frac{\partial \vec{A}}{\partial t} \sim \frac{1}{r^2}$,
 $da \sim r^2$

$$\therefore \oint_S \frac{1}{\mu_0} \vec{B} \times \frac{\partial \vec{A}}{\partial t} \cdot d\vec{a} \rightarrow 0$$

$$\therefore \frac{dW_m}{dt} = \frac{1}{\mu_0} \int \vec{B} \cdot \frac{\partial \vec{B}}{\partial t} \, d\tau$$

$$= \frac{1}{\mu_0} \int \frac{\partial}{\partial t} \left(\frac{1}{2} \vec{B} \cdot \vec{B} \right) d\tau$$

$$= \frac{d}{dt} \int \frac{1}{2\mu_0} B^2 \, d\tau \quad \dots (33)$$

By setting $W_m = 0$ when $B = 0$

$$\therefore W_m = \frac{1}{2\mu_0} \int B^2 \, d\tau \quad \dots (34)$$

Eq. (34) has a similar form to

the electrostatic energy $\int \frac{1}{2} \epsilon_0 E^2 dz$.

When there are magnetic materials,

$$\vec{J}_f = \vec{\nabla} \times \vec{H}$$

One gets
$$\frac{dW_m}{dt} = \int \frac{\partial \vec{A}}{\partial t} \cdot \vec{\nabla} \times \vec{H} dz$$

$$= \int \vec{H} \cdot \vec{\nabla} \times \frac{\partial \vec{A}}{\partial t} dz$$

$$= \int \vec{H} \cdot \frac{\partial}{\partial t} \vec{B} dz \quad \dots (35)$$

Hence
$$\Delta W_m = \int \vec{H} \cdot \Delta \vec{B} dz \quad \dots (36)$$

$$W_m = \int dz \int_0^B \vec{H} \cdot \Delta \vec{B} \quad \dots (36) - 1$$

For linear media, $\vec{H} = \frac{1}{\mu} \vec{B}$,


One gets
$$\frac{dW_m}{dt} = \int \frac{1}{\mu} \vec{B} \cdot \frac{\partial \vec{B}}{\partial t} dz$$

$$= \frac{d}{dt} \int \frac{1}{2\mu} B^2 dz$$

$$\therefore W_m = \int \frac{1}{2\mu(\vec{r})} \vec{B}^2(\vec{r}, t) dz \quad \dots (37)$$

which has a similar form to

$$W_E = \int \frac{1}{2} \epsilon E^2 dz$$

\therefore Eq. (37) includes the energy that generates  Dr. J. C. Das

Other forms

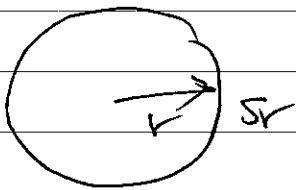
Eq. (34) can be converted into other forms.

Using $\vec{B} = \vec{\nabla} \times \vec{A}$, one has

$$\vec{B}^2 = \vec{B} \cdot \vec{\nabla} \times \vec{A} = \vec{\nabla} \cdot (\vec{A} \times \vec{B}) + \vec{A} \cdot \vec{\nabla} \times \vec{B}$$

$$\therefore W_m = \frac{1}{2\mu_0} \int [\vec{\nabla} \cdot (\vec{A} \times \vec{B}) + \vec{A} \cdot \vec{\nabla} \times \vec{B}] dz$$

$$\int \vec{\nabla} \cdot (\vec{A} \times \vec{B}) dz = \int_{S_r} \vec{A} \times \vec{B} \cdot d\vec{a} \rightarrow 0 \text{ as } r \rightarrow \infty$$



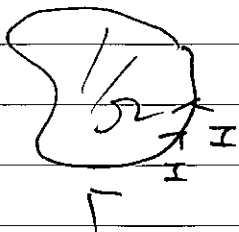
$$\therefore W_m = \frac{1}{2\mu_0} \int \vec{A} \cdot \vec{\nabla} \times \vec{B} dz$$

$$= \frac{1}{2} \int \vec{A} \cdot \vec{J}_A dz$$

L -- (38)

When ^{the} current is in a circuit, one gets

$$W_m = \frac{1}{2} \oint_{\Gamma} \vec{A} \cdot d\vec{l} I$$



$$\text{Now } \oint \vec{A} \cdot d\vec{l} = \int_{\Sigma} \nabla \times \vec{A} \cdot d\vec{a}$$

$$= \Phi_B = LI$$

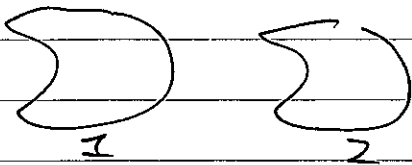
L = self inductance of loop Γ .

$$\therefore W_m = \frac{1}{2} LI^2, \quad \frac{dW_m}{dt} = LI \frac{dI}{dt} = \mathcal{E}I$$

L -- (39)

In general, if there are more than two

Circuits, say, Γ_1 & Γ_2 ,



$$W_m = \sum_n \frac{1}{2} \oint_{\Gamma_n} \vec{A} \cdot d\vec{s} I_n$$

$$\oint_{\Gamma_n} \vec{A} \cdot d\vec{s} = \Phi_n = \sum_m \Phi_{nm} = L_n I_n + \sum_{m \neq n} M_{nm} I_m$$

$$\therefore W_m = \sum_n \frac{1}{2} L_n I_n^2 + \sum_{\substack{n, m \\ n \neq m}} \frac{1}{2} I_n M_{nm} I_m \quad \dots (40)$$

where L_n is the inductance of loop n .

and M_{nm} is the mutual inductance between loop n & m .

For two loops, one gets

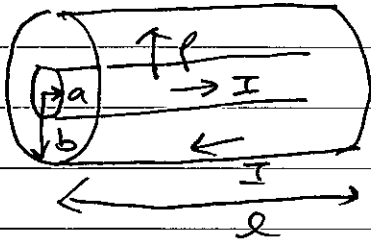
$$\oint_{\Gamma_1} \vec{A} \cdot d\vec{s} = L_1 I_1 + M I_2$$

$$\oint_{\Gamma_2} \vec{A} \cdot d\vec{s} = L_2 I_2 + M I_1$$

$$\begin{aligned} \therefore W_m &= \frac{1}{2} L_1 I_1^2 + \frac{1}{2} L_2 I_2^2 + \frac{1}{2} I_1 M I_2 + \frac{1}{2} I_2 M I_1 \\ &= \frac{1}{2} L_1 I_1^2 + \frac{1}{2} L_2 I_2^2 + M I_1 I_2 \end{aligned}$$

Example. A long coaxial cable carries

current I as shown in the left figure. Find the self inductance.



Solution: L can be found by calculating the energy $W_m = \frac{1}{2} L I^2$

According to Ampere's law,

$$\vec{B} = \frac{\mu_0 I}{2\pi r} \hat{\phi} \quad \text{for } a < r < b$$

$$\therefore \frac{1}{2\mu_0} B^2 = \frac{\mu_0}{8\pi^2} \frac{I^2}{r^2}$$

$$\therefore dz = 2\pi r dr \cdot l$$

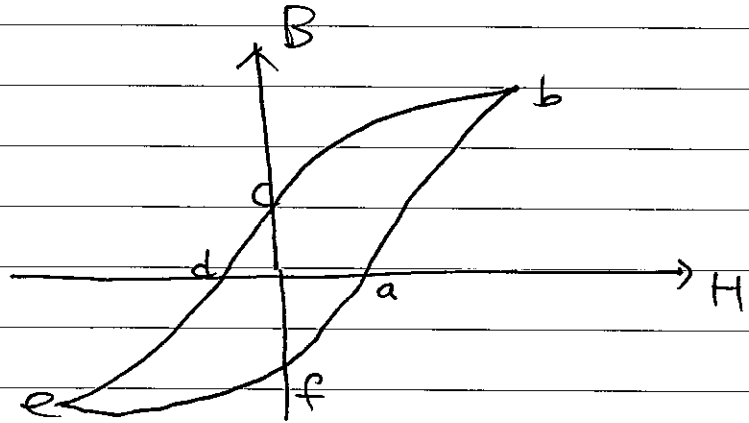
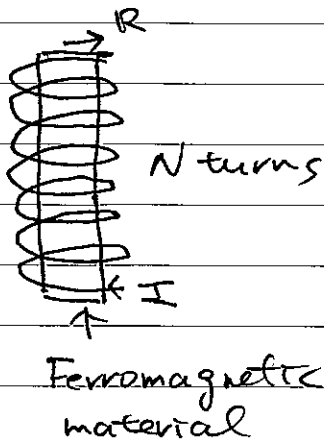
$$\therefore W_m = \int \frac{1}{2\mu_0} B^2 dz$$

$$= 2\pi l \cdot \frac{\mu_0}{8\pi^2} I^2 \int_a^b \frac{r dr}{r^2}$$

$$= \frac{\mu_0 I^2}{4\pi} l \ln \frac{b}{a} = \frac{1}{2} L I^2$$

$$\therefore L = \frac{\mu_0 l}{2\pi} \ln \frac{b}{a}$$

Example. Energy dissipation in a hysteresis cycle



Solution: By increasing \$I\$, \$H \uparrow\$, one

polarizes \$M\$ inside the coil. One needs

to do work against the opposed emf (\$-N \frac{d\Phi}{dt}\$)

due to Lenz's law

$$\therefore \frac{dW}{dt} = I N \frac{d\Phi}{dt}, \quad \Phi = \pi R^2 B$$

$$\therefore \frac{dW}{dt} = \frac{IN}{l} \cdot \frac{\pi R^2}{V} \frac{dB}{dt}$$

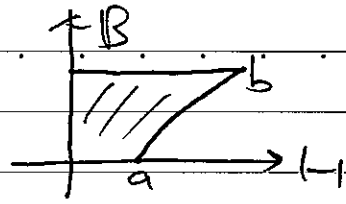
$\underbrace{IN}_{H=lI} \quad \underbrace{\frac{\pi R^2}{V}}_1$

$$\therefore \frac{dW}{dt} = V H \frac{dB}{dt}$$

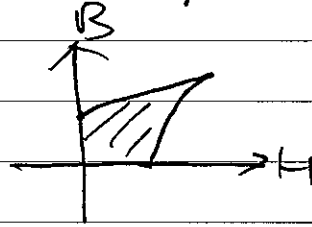
$$W = V \int H dB \quad \text{is consistent with}$$

eg. (36)

$\therefore W_{a \rightarrow b} = V \cdot$ Shaded area



$W_{b \rightarrow c} = V \cdot$ area of



\therefore total work = $V \cdot \oint H dB$

= $V \cdot$ area of



\therefore For each cycle, one needs to do the work is proportional to the area of hysteresis loop.

Maxwell's equations

Inconsistency of Ampere's law &

other laws

Up to now, we have obtained 4 equations for \vec{E} & \vec{B} :

(i) $\vec{\nabla} \cdot \vec{E} = \rho / \epsilon_0$ (Gauss's law)

(ii) $\vec{\nabla} \cdot \vec{B} = 0$

(iii) $\vec{\nabla} \times \vec{E} = -\frac{d\vec{B}}{dt}$ (Faraday's law)

(iv) $\vec{\nabla} \times \vec{B} = \mu_0 \vec{J}$ (Ampere's law).

(ii) & (iii) are consistent as one can check.

$$\begin{aligned} \vec{\nabla} \cdot (\vec{\nabla} \times \vec{E}) &= -\vec{\nabla} \cdot \frac{d\vec{B}}{dt} \\ &\stackrel{||}{=} 0 \qquad = -\frac{d}{dt} \underbrace{\vec{\nabla} \cdot \vec{B}}_{\stackrel{||}{=} 0 \text{ (iii)}} = 0 \end{aligned}$$

$\therefore 0=0$ (ii) & (iii) are consistent.

However, $\vec{\nabla} \times \vec{B} = \mu_0 \vec{J}$ is only valid

$$\text{When } \vec{\nabla} \cdot \vec{J} = 0 : \quad \underbrace{\vec{\nabla} \cdot (\vec{\nabla} \times \vec{B})}_{||=0} = \mu_0 \vec{\nabla} \cdot \vec{J} \quad \text{--- (4)}$$

$\therefore \vec{\nabla} \cdot \vec{J}$ must vanish.

From the conservation of charge,

$$\vec{\nabla} \cdot \vec{J} = -\frac{d\rho}{dt} \quad \text{which does not}$$

generally vanish!

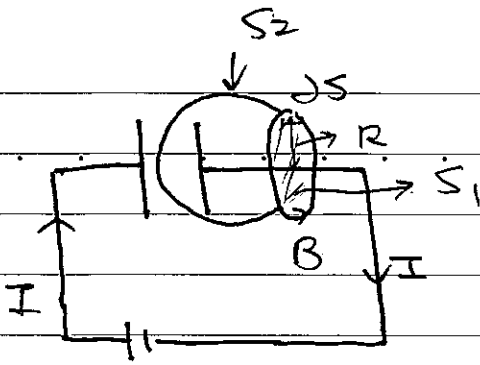
Hence Ampere's law is not consistent with the continuity equation.

Another way to see that the Ampere's law

is not complete is to consider the

charging of a capacitor as shown

in the below.



Applying the Ampere's law to dS , one

gets

$$\oint_{dS} \vec{B} \cdot d\vec{e} = \mu_0 I_{\text{enclosed}} = \mu_0 I$$

L. (42)

However, using the Stokes's theorem,

one can rewrite

$$\oint_{dS} \vec{B} \cdot d\vec{e} = \int_{S_1} \vec{\nabla} \times \vec{B} \cdot d\vec{a} \quad \dots (43)$$

where dS is the boundary of S_1 .

There are many ways to choose S_1 as long as it is bounded by dS .

$\therefore S_2$ is equally correct:

$$\therefore \oint_{dS} \vec{B} \cdot d\vec{e} = \int_{S_2} \vec{\nabla} \times \vec{B} \cdot d\vec{a}$$

$$\stackrel{(iv)}{\Rightarrow} \int_{S_2} \mu_0 \vec{J} \cdot d\vec{a}$$

but $\vec{J} = 0$ on S_2 ! One gets a conflict result $\oint_{dS} \vec{B} \cdot d\vec{e} = 0$!

Fixed Ampere's Law

Maxwell fixed the conflict by noting that

$$\begin{aligned}\vec{\nabla} \cdot \vec{J} &= -\frac{d\rho}{dt} = -\frac{d}{dt}(\epsilon_0 \vec{\nabla} \cdot \vec{E}) \\ &= -\vec{\nabla} \cdot \epsilon_0 \frac{d\vec{E}}{dt}\end{aligned}$$

Hence, if one adds $\mu_0 \epsilon_0 \frac{d\vec{E}}{dt}$ to Right hand side of (iv), eq. (4) becomes

$$\vec{\nabla} \cdot (\vec{\nabla} \times \vec{B}) = \mu_0 \vec{\nabla} \cdot \vec{J} + \mu_0 \epsilon_0 \frac{d\vec{\nabla} \cdot \vec{E}}{dt}$$

both LHS & RHS Vanish now!

$$\therefore \vec{\nabla} \times \vec{B} = \mu_0 \vec{J} + \mu_0 \epsilon_0 \frac{d\vec{E}}{dt} \quad \dots (44)$$

can resolve the conflict shown in eq. (4)

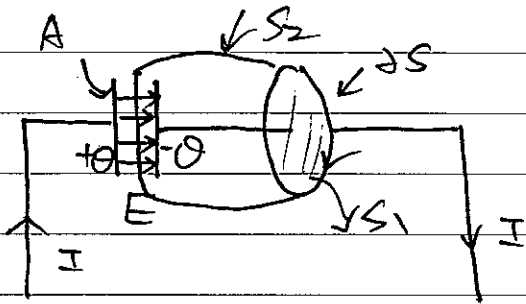
\therefore Changing electric field contributes a current and induces magnetic field!

The term $\epsilon_0 \frac{d\vec{E}}{dt}$ is called the displacement current

$$\vec{J}_D = \epsilon_0 \frac{d\vec{E}}{dt} \quad \dots (45)$$

The inclusion of \vec{J}_D also resolves the conflict.

of charging a capacitor. ∴



$$E = \frac{\sigma}{\epsilon_0} = \frac{Q}{\epsilon_0 A}$$

$$\therefore \epsilon_0 \frac{dE}{dt} = \frac{1}{A} \frac{dQ}{dt} = \frac{I}{A}$$

$$\therefore \oint_{S_2} \vec{B} \cdot d\vec{\ell} = \oint_{S_2} \vec{\nabla} \times \vec{B} \cdot d\vec{a}$$

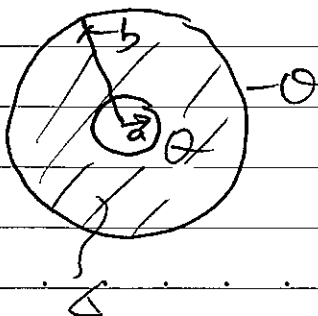
$$= \oint \underbrace{\mu_0 \epsilon_0 \frac{d\vec{E}}{dt}}_{I/A} \cdot d\vec{a}$$

$$= \mu_0 I / A \times A$$

Which is now consistent with the calculation of S_1 .

Example

Find \vec{J} & \vec{B} in $a < r < b$.



$$\vec{J} = \sigma \vec{E} = \sigma \frac{Q}{4\pi \epsilon_0 r^2} \hat{r}$$

$$I = \int \vec{J} \cdot d\vec{a} = \frac{\sigma Q}{\epsilon_0} \quad \text{--- (4)}$$

By symmetry and $\vec{\nabla} \cdot \vec{B} = 0$, $\oint \vec{B} \cdot d\vec{a} = 0$

$$\therefore B \cdot 4\pi r^2 = 0 \quad B = 0$$

Why I doesn't generate B?

Solution : $\vec{J}_d = \epsilon_0 \frac{d\vec{E}}{dt} = \frac{1}{4\pi r^2} \frac{dQ}{dt} \hat{r}$

$$= \frac{1}{4\pi r^2} \left(-\frac{dQ}{dt} \right) \hat{r}$$

↑
ef. ⊗

$$\therefore \vec{J} + \vec{J}_d = 0 \Rightarrow \vec{\nabla} \times \vec{B} = 0$$

together with $\vec{\nabla} \cdot \vec{B} = 0$

$$\therefore \vec{B} = 0$$

Maxwell equations

After the important work by Maxwell, the equations that characterize \vec{E} & \vec{B}

in most general situations are

the (i) $\vec{\nabla} \cdot \vec{E} = \rho / \epsilon_0$ Gauss's law

(ii) $\vec{\nabla} \cdot \vec{B} = 0$

(iii) $\vec{\nabla} \times \vec{E} = -\frac{d\vec{B}}{dt}$ Faraday's law

(iv) $\vec{\nabla} \times \vec{B} = \mu_0 \left(\vec{J} + \epsilon_0 \frac{d\vec{E}}{dt} \right)$ Ampere-Maxwell law

These equations are now complete and

called Maxwell's equations.

Together with $\vec{F} = q(\vec{E} + \vec{v} \times \vec{B})$,

they describe the entire classical electrodynamics.

Note that $\vec{\nabla} \cdot \vec{J} = -\frac{d\rho}{dt}$ is a consequence
 \uparrow
 the continuity equation

of eq. (IV) if one takes $\vec{\nabla} \cdot$ (IV).

In the above form, it's tempting to

conclude that \vec{E} can be produced by either ρ or $-\frac{d\vec{B}}{dt}$ and \vec{B} can be produced either by \vec{J} or $\frac{d\vec{E}}{dt}$.

This, however, is quite correct as all \vec{E} & \vec{B} are produced by ρ & \vec{J} .

Hence it's more logical to write

(i) $\vec{\nabla} \cdot \vec{E} = \rho/\epsilon_0$ (ii) $\vec{\nabla} \times \vec{E} + \frac{d\vec{B}}{dt} = 0$

(iii) $\vec{\nabla} \cdot \vec{B} = 0$ (IV) $\vec{\nabla} \times \vec{B} - \mu_0 \epsilon_0 \frac{d\vec{E}}{dt} = \mu_0 \vec{J}$ (47)

So that all sources are on the right hand side.

Magnetic monopole

The Maxwell equations are symmetric

in \vec{E} & \vec{B} in the absence of matters:

$$\vec{\nabla} \cdot \vec{E} = 0 \quad (+\rho/\epsilon_0) \quad \vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$$

$$\vec{\nabla} \cdot \vec{B} = 0 \quad \vec{\nabla} \times \vec{B} = \mu_0 \epsilon_0 \frac{\partial \vec{E}}{\partial t} \quad (+\mu_0 \vec{J})$$

change (\vec{E}, \vec{B}) by $(\vec{B}, -\epsilon_0 \mu_0 \vec{E})$

exchange the first pair of equations

into the second pair of equations.

This symmetry is spoiled by matters

when ρ & \vec{J} are present!

The spoil of the symmetry between \vec{E} & \vec{B}

is easily traced to be due to the absence of magnetic

charges (monopoles).

If we had

$$(i) \vec{\nabla} \cdot \vec{E} = \frac{\rho_e}{\epsilon_0} \quad (iii) \vec{\nabla} \times \vec{E} = -\mu_0 \vec{J}_m - \frac{d\vec{B}}{dt}$$

$$(ii) \vec{\nabla} \cdot \vec{B} = \mu_0 \rho_m \quad (iv) \vec{\nabla} \times \vec{B} = \mu_0 \vec{J}_e + \mu_0 \epsilon_0 \frac{d\vec{E}}{dt}$$

L- (48)

With ρ_m = density of magnetic charges

\vec{J}_m = current of magnetic charges

Both charges are conserved

$$\vec{\nabla} \cdot \vec{J}_m = -\frac{d\rho_m}{dt}, \quad \vec{\nabla} \cdot \vec{J}_e = -\frac{d\rho_e}{dt}$$

L- (49)

$$\text{then } (\vec{E}, \vec{B}) \rightarrow \left(\frac{\vec{B}}{\sqrt{\mu_0 \epsilon_0}}, \sqrt{\mu_0 \epsilon_0} \vec{E} \right)$$

$$(\rho_e, \rho_m) \rightarrow (\sqrt{\epsilon_0 \mu_0} \rho_m, -\frac{1}{\sqrt{\mu_0 \epsilon_0}} \rho_e)$$

$$(\vec{J}_e, \vec{J}_m) \rightarrow (\sqrt{\epsilon_0 \mu_0} \vec{J}_m, -\frac{1}{\sqrt{\mu_0 \epsilon_0}} \vec{J}_e)$$

Eqs (48) are invariant.

In fact, there are more transformations that leave eq. (48) invariant. (problem 7.64)

The Maxwell equations seem to be

more complete if there are magnetic charges.

In spite of this, however, up to now, no one

has ever found any magnetic monopole.

Maxwell's equations in Matter

The Maxwell's equations, eq. (47), are

complete. However, inside materials, it's

more convenient to separate responses of

materials \vec{P} & \vec{J} and express everything

in terms of \vec{P} & \vec{J} .

For static case, we have

$$\rho_b = -\vec{\nabla} \cdot \vec{P} \quad (\text{bound charge})$$

$$\vec{J}_b = \vec{\nabla} \times \vec{M} \quad (\text{bound current}).$$

When we consider non-static situations,

changes in ρ_b & \vec{J}_b will result in ^a new

quantity. ∴

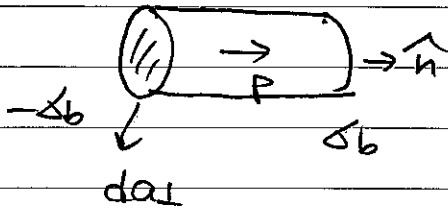
Polarization current \vec{J}_p

Change of \vec{P} in time results in

the polarization current $\vec{J}_p = \frac{d\vec{P}}{dt} \dots (50)$

This can be understood by considering

a small volume with \vec{P} . Clearly, there are



$\pm \Delta b$ at two ends

$$\text{with } \Delta b = \vec{P} \cdot \hat{n} = P$$

When P increases, Δb increases. Therefore,

there are charges moving from the left end to the right end

The current
$$I = \frac{dQ}{dt} = \frac{d}{dt} (\Delta b \cdot da_L)$$

$$= \frac{dP}{dt} \cdot da_L$$

$$\therefore \vec{J}_p = \frac{d\vec{P}}{dt}$$

Microscopically, when \vec{P} increases,

$+$ moves right, $-$ moves left, therefore

there is a net move of charges to the right. This is the origin of \vec{J}_p .

Continuity equation:
$$\vec{\nabla} \cdot \vec{J}_p = -\vec{\nabla} \cdot \frac{d\vec{P}}{dt}$$

$$= -\frac{d}{dt} \vec{\nabla} \cdot \vec{P} = -\frac{d\rho_b}{dt}$$

$$\therefore \frac{dP_b}{dt} + \vec{\nabla} \cdot \vec{J}_p = 0 \quad \dots (51)$$

With \vec{J}_p , the general decompositions

of P & \vec{J} are

$$P = P_f + P_b = P_f - \vec{\nabla} \cdot \vec{P} \quad \dots (52)$$

$$\vec{J} = \vec{J}_f + \vec{J}_b + \vec{J}_p$$

$$= \vec{J}_f + \vec{\nabla} \times \vec{M} + \frac{d\vec{P}}{dt} \quad \dots (53)$$

\therefore Gauss's law

$$\vec{\nabla} \cdot \vec{E} = \frac{\rho}{\epsilon_0} = \frac{P_f}{\epsilon_0} - \frac{1}{\epsilon_0} \vec{\nabla} \cdot \vec{P}$$

becomes

$$\vec{\nabla} \cdot \vec{D} = P_f$$

$$\text{with } \vec{D} \equiv \epsilon_0 \vec{E} + \vec{P}$$

While for the Ampere's law,

$$\vec{\nabla} \times \vec{B} = \mu_0 \vec{J} + \mu_0 \epsilon_0 \frac{d\vec{E}}{dt}$$

$$= \mu_0 (\vec{J}_f + \vec{\nabla} \times \vec{M} + \frac{d\vec{P}}{dt}) + \mu_0 \epsilon_0 \frac{d\vec{E}}{dt}$$

$$\Rightarrow \vec{\nabla} \times \left(\frac{\vec{B}}{\mu_0} - \vec{M} \right) = \vec{J}_f + \frac{d}{dt} (\epsilon_0 \vec{E} + \vec{P})$$

$$\therefore \vec{\nabla} \times \vec{H} = \vec{J}_f + \frac{d\vec{D}}{dt}$$

$$\text{with } \vec{H} \equiv \frac{\vec{B}}{\mu_0} - \vec{M}$$

Hence, in terms of ρ_f & \vec{J}_f , the Maxwell's

equations become

$$(i) \vec{D} \cdot \vec{D} = \rho_f$$

$$(ii) \vec{D} \times \vec{E} = -\frac{d\vec{B}}{dt}$$

$$(iii) \vec{D} \cdot \vec{B} = 0$$

$$(iv) \vec{D} \times \vec{H} = \vec{J}_f + \frac{d\vec{D}}{dt}$$

which have to be supplemented by \dots (54)

the constitution relations :

$$\vec{D} = \epsilon \vec{E}, \quad \vec{H} = \vec{B} / \mu$$

$$\vec{P} = \epsilon_0 \chi_e \vec{E}, \quad \vec{M} = \chi_m \vec{H}$$

Where $\epsilon = \epsilon_0 (1 + \chi_e)$, $\mu = \mu_0 (1 + \chi_m)$

In this context, $\vec{J}_d = \frac{d\vec{D}}{dt}$ is the displacement current.

Boundary conditions

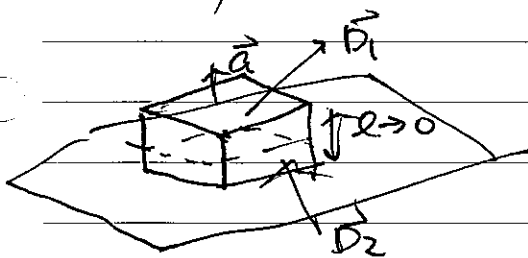
In the presence of surface charge density

& \vec{J}_f current density \vec{K} , \vec{E} , \vec{B} , \vec{D} & \vec{H}

may be discontinuous.

(i) using a pillbox near surface with $l \rightarrow 0$

$$\oint \vec{D} \cdot d\vec{a} = \vec{D}_1 \cdot \vec{a} - \vec{D}_2 \cdot \vec{a} = \sigma_f a$$

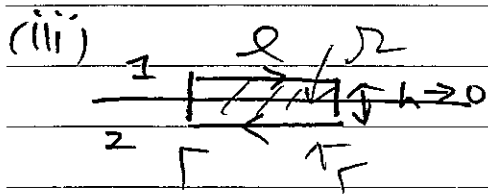


$$\vec{D} \cdot \vec{a} = D^{\perp} a$$

$$\therefore D_1^{\perp} - D_2^{\perp} = \Delta f \quad \dots (55)$$

(ii) Similarly, $\because \vec{D} \cdot \vec{B} = 0$, using the same

pillar box, one gets $B_1^{\perp} - B_2^{\perp} = 0 \quad \dots (56)$



Consider an Amperian loop

shown in the left figure.

$h \rightarrow 0$

$$\oint \vec{E} \cdot d\vec{e} = \int_{\Omega} \nabla \times \vec{E} \cdot d\vec{a}$$

$$= \int_{\Omega} -\frac{d\vec{B}}{dt} \cdot d\vec{a} \rightarrow 0 \text{ as } h \rightarrow 0$$

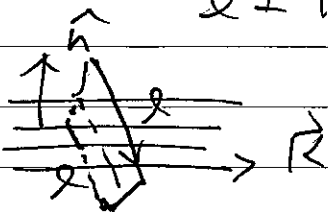
& $\frac{d\vec{B}}{dt}$ is finite

$$\text{But } \oint \vec{E} \cdot d\vec{e} = \vec{E}_1 \cdot \vec{l} - \vec{E}_2 \cdot \vec{l} \quad (h \rightarrow 0)$$

$$\therefore E_1^{\parallel} - E_2^{\parallel} = 0 \quad \dots (57)$$

(iv) Using the same loop as (iii) but with

$\vec{D} \perp \vec{R}$



$$\oint \vec{H} \cdot d\vec{e} = \int \nabla \times \vec{H} \cdot d\vec{a}$$

$$= \int (\vec{J}_f + \frac{d\vec{D}}{dt}) \cdot d\vec{a}$$

$$= \vec{K}_f \cdot \vec{n} \times \vec{e}$$

$$\vec{K} \times \hat{n} \cdot \vec{\ell}$$

∥

$$\therefore \vec{H}_1 \cdot \vec{\ell} - \vec{H}_2 \cdot \vec{\ell} = \vec{K} \cdot (\hat{n} \times \vec{\ell}) \quad \text{for any } \vec{\ell}$$

$$\therefore \vec{H}_1 - \vec{H}_2 = \vec{K} \times \hat{n} \quad \dots \textcircled{5D}$$

Eqs. $\textcircled{5J}$ - $\textcircled{5F}$ are the general boundary conditions for electrodynamics with materials.

For linear media, one can express them in terms of \vec{E} & \vec{B} :

$$(i) \epsilon_1 E_1^\perp - \epsilon_2 E_2^\perp = \sigma_f$$

$$(ii) \vec{E}_1^\parallel - \vec{E}_2^\parallel = 0$$

$$(iii) B_1^\perp - B_2^\perp = 0$$

$$(iv) \frac{1}{\mu_1} \vec{B}_1^\parallel - \frac{1}{\mu_2} \vec{B}_2^\parallel = \vec{K} \times \hat{n}$$

These equations will be the basis for the theory of electrodynamics in the materials.

Quasi-static approximations of AC circuit theory

Slowly time-varying charges &

quasi-electrostatic approximation

In the following, we shall use l to characterize the size of the system considered ^{typical}

$\therefore \nabla \sim 1/l$, and use ω to characterize the typical time scale $\therefore \omega \sim 1/T$.

If the EM fields are generated primarily by a slowly time-changing charge

density $\rho(\vec{r}, t)$, one can write

$$\vec{E} = \vec{E}_C + \vec{E}_{in} \quad \text{with} \quad \vec{\nabla} \times \vec{E}_C = 0, \quad \vec{\nabla} \cdot \vec{E}_{in} = 0$$

[↑] "Coulomb" [↑] "Faraday", \vec{E} is primarily \vec{E}_C

$$\therefore \vec{\nabla} \cdot \vec{E} = \vec{\nabla} \cdot \vec{E}_C = \rho / \epsilon_0, \quad E_C/l \sim \rho / \epsilon_0, \quad E_C \sim \frac{l}{\epsilon_0} \rho$$

$$\vec{\nabla} \times \vec{E}_{in} = -\frac{d\vec{B}}{dt}, \quad E_{in}/l \sim \omega B, \quad E_{in} \sim \omega B l$$

$$\therefore \vec{\nabla} \cdot \vec{J} + \frac{d\rho}{dt} = 0 \quad \therefore J/l \sim \omega \rho, \quad J \sim \omega l \rho$$

$$J_0 = \epsilon_0 \frac{dE_C}{dt} + \epsilon_0 \frac{dE_{in}}{dt} \sim \epsilon_0 \omega (E_C + E_{in})$$

$$\therefore J_0 \text{ is at least } \epsilon_0 \omega E_C, \quad J_0 \geq \epsilon_0 \omega E_C$$

$$\therefore \frac{J}{J_0} \geq \frac{\omega l \rho}{\epsilon_0 \omega E_C} \sim \frac{\omega l \rho}{\rho l \omega} \sim 1$$

From $\vec{\nabla} \times \vec{B} = \mu_0 (\vec{J} + \vec{J}_0) \sim \mu_0 \vec{J}$

$$\therefore B/l \sim \mu_0 J$$

$$B \sim \mu_0 J l \sim \mu_0 \omega \rho l^2 \quad \dots (5)$$

Hence $E_{in} \sim \omega B l \sim \mu_0 \omega^2 \rho l^3$

$$\frac{E_{in}}{E_c} \sim \frac{\mu_0 \omega^2 \rho l^3}{\rho / \epsilon_0} \sim \epsilon_0 \mu_0 \omega^2 l^2$$

$$= \frac{\omega^2 l^2}{c^2} \ll 1 \quad \dots (6)$$

if $l \ll c/\omega$ or $\omega \ll c/l$

This is quasi-electrostatic approximation. One

has

$$\vec{\nabla} \cdot \vec{E} = \rho / \epsilon_0 \quad \vec{\nabla} \cdot \vec{B} = 0$$

$$\vec{\nabla} \times \vec{E} = 0 \quad \vec{\nabla} \times \vec{B} = \mu_0 \left(\vec{J} + \epsilon_0 \frac{d\vec{E}}{dt} \right)$$

(E_F can be neglected. $\vec{\nabla} \times \vec{E} = 0$)

$$\therefore \vec{\nabla} \times \vec{E} = 0 \quad \therefore \vec{B} = \vec{\nabla} \times \vec{A} = \frac{\mu_0}{4\pi} \int \frac{\vec{J}(\vec{r}') \times (\vec{r} - \vec{r}')}{|\vec{r} - \vec{r}'|^3} dz'$$

($\vec{\nabla} \times \vec{J}_0 = 0$)

$$\vec{E} = -\nabla \frac{1}{4\pi\epsilon_0} \int \frac{\rho(\vec{r}')}{|\vec{r} - \vec{r}'|} dz'$$

$$\vec{A} = \frac{\mu_0}{4\pi} \int \frac{\vec{J}(\vec{r}')}{|\vec{r} - \vec{r}'|} dz' \quad \dots (6)$$

In this case, taking the static limit, one gets electrostatic

Slowly time-varying currents &

quasi-magnetostatic approximation

If the EM fields are generated by

slowly time-changing currents $\vec{J}(\vec{r}, t)$,

the analysis is different. (\because

ρ is induced by $\vec{\nabla} \cdot \vec{J}$.)

In this case, we start by estimating

the magnetic field generated by \vec{J} . via

the Ampere's law.

$$\vec{\nabla} \times \vec{B}_A = \mu_0 \vec{J}$$

$$\therefore B_A \sim \mu_0 J l$$

which in turn generates

$$\vec{\nabla} \times \vec{E}_{in} = -\frac{d\vec{B}_A}{dt}$$

$$\therefore E_{in} \sim \omega B_A \sim \mu_0 \omega l^2 J$$

$$\therefore \vec{J}_D = \epsilon_0 \frac{d\vec{E}_{in}}{dt} \quad J_D \sim \epsilon_0 \omega E_{in} \sim \epsilon_0 \omega \mu_0 \omega l^2 J$$

$$\frac{J_D}{J} \sim \mu_0 \epsilon_0 \omega^2 l^2 = \frac{\omega^2 l^2}{c^2} \ll 1 \quad (62)$$

if $l \ll c/\omega$ or $\omega \ll c/l$

This is the quasi-magnetostatic approximation in which J_0 is neglected.

$$\vec{\nabla} \cdot \vec{E} = \rho / \epsilon_0, \quad \vec{\nabla} \cdot \vec{B} = 0$$

$$\vec{\nabla} \times \vec{E} = -\frac{d\vec{B}}{dt}, \quad \vec{\nabla} \times \vec{B} = \mu_0 \vec{J}$$

$$\therefore \vec{\nabla} \cdot (\vec{\nabla} \times \vec{B}) = 0 \quad \therefore \vec{\nabla} \cdot \vec{J} = 0$$

$\rho = \rho(\vec{r})$ one can choose $\rho = 0$

So that in the static limit, one recovers the magnetostatics, $\vec{\nabla} \cdot \vec{E} = 0$ $\vec{\nabla} \cdot \vec{B} = 0$

$$\vec{\nabla} \times \vec{E} = -\frac{d\vec{B}}{dt} \quad \vec{\nabla} \times \vec{B} = \mu_0 \vec{J}$$

In the quasi-magnetostatic approximation,

$$\vec{B}(\vec{r}, t) = \vec{\nabla} \times \vec{A}$$

$$= \vec{\nabla} \times \left(\frac{\mu_0}{4\pi} \int \frac{\vec{J}(\vec{r}', t')}{|\vec{r} - \vec{r}'|} dz' \right)$$

$$\therefore \vec{E}(\vec{r}, t) = -\frac{\mu_0}{4\pi} \int \frac{\frac{d\vec{J}(\vec{r}', t')}{dt'}}{|\vec{r} - \vec{r}'|} dz'$$

$$= -\frac{d\vec{A}}{dt}$$

--- (63)

In general, when both charges and currents are slowly-changing, it is the quasi-static limit. (or quasi-stationary)

In this case, when $l \ll c\tau$

(or $\omega \ll c/l$, $l \ll \lambda$)

(i) both \vec{J} & \vec{J}_0 are present,

one can neglect $\vec{J}_0 \therefore \vec{\nabla} \times \vec{B} = \mu_0 \vec{J}$
 $\vec{\nabla} \cdot \vec{B} = 0$

(ii) both \vec{E}_m and \vec{E}_c are present

one can neglect \vec{E}_m

$\therefore \vec{\nabla} \cdot \vec{E} = \rho/\epsilon_0$, $\vec{\nabla} \times \vec{E} = 0$

\therefore There is no direct coupling between \vec{E} and \vec{B} in quasi-static approximation.

The quasi-static approximation underlies the most application of alternating-current (AC)

Circuit when sizes of circuit $l \ll \lambda$

with $\omega =$ frequency of AC

Quasi-static approximation for good conductors.

Charge relaxation in conductors

Inside a conductor, $\vec{J}_f = \sigma \vec{E}$

$$\therefore \nabla \cdot \vec{E} = \rho_f$$

$$\therefore \nabla \cdot \vec{J}_f = \sigma \nabla \cdot \vec{E} = \frac{\sigma}{\epsilon} \rho_f \quad \dots (64)$$

The continuity equation $\nabla \cdot \vec{J}_f + \frac{d\rho_f}{dt} = 0$

becomes
$$\frac{d\rho_f}{dt} + \frac{\sigma}{\epsilon} \rho_f = 0 \quad \dots (65)$$

$$\therefore \rho_f(\vec{r}, t) = \rho_f(\vec{r}, 0) e^{-t/\tau_E} \quad \dots (66)$$

$$\tau_E = \epsilon/\sigma$$

$$\therefore \text{for } t \gg \tau_E = \epsilon/\sigma, \quad \rho_f = 0$$

$$\therefore \nabla \cdot \vec{J}_f = 0 \quad \dots (67)$$

$$\therefore t \sim 1/\omega \quad \therefore \text{For } \omega \ll \frac{\sigma}{\epsilon}, \quad \nabla \cdot \vec{J}_f = 0$$

($\omega \tau_E \ll 1$)

Now,
$$\vec{J}_D = \epsilon \frac{d\vec{E}}{dt} \sim \epsilon \omega \vec{E}$$

$$\vec{J}_f = \sigma \vec{E}$$

$$\therefore \frac{J_D}{J_f} = \frac{\epsilon \omega}{\sigma} = \omega \tau_E \quad \dots (68)$$

Similar to quasi-magnetostatic approximation,

In principle, there are \vec{J}_{ext} due to external sources. The induced \vec{J}_0 due to \vec{J}_{ext}

satisfies $\frac{\vec{J}_0}{\vec{J}_{ext}} \sim \frac{\omega^2 \epsilon^2}{c^2} \dots (6P)$

(cf. eq. (62))

Hence for $\omega \ll c/\lambda$ & $\omega \ll \sigma/\omega$, one

can ignore \vec{J}_0 .

In the absence of \vec{J}_{ext} , one gets

The quasi-static approximation for good conductors:

$$\vec{\nabla} \cdot \vec{E} = 0 \quad \vec{\nabla} \times \vec{E} = -\frac{d\vec{B}}{dt}$$

$$\vec{\nabla} \cdot \vec{B} = 0 \quad \vec{\nabla} \times \vec{B} = \mu \sigma \vec{E}$$

By taking $\vec{\nabla} \times (\vec{\nabla} \times \vec{E})$, one gets

$$\nabla^2 \vec{E} = \mu \sigma \frac{d\vec{E}}{dt}$$

Similarly, taking $\vec{\nabla} \times (\vec{\nabla} \times \vec{B})$, one gets

$$\nabla^2 \vec{B} = \mu \sigma \frac{d\vec{B}}{dt}$$

as well. Hence both \vec{E} & \vec{B} satisfy

The diffusion equation: $\frac{d}{dt} \psi = D \nabla^2 \psi$, $D = \frac{1}{\mu \sigma}$

$\psi = \vec{E}$ or \vec{B} , $\dots (70)$

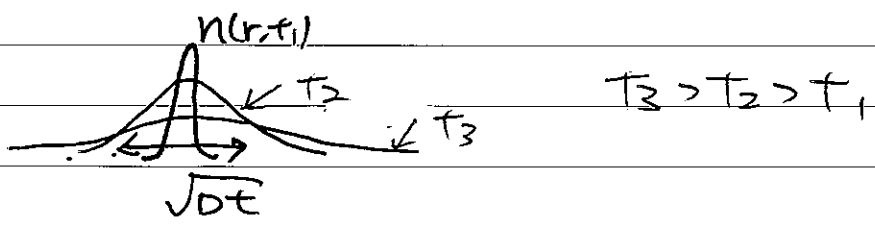
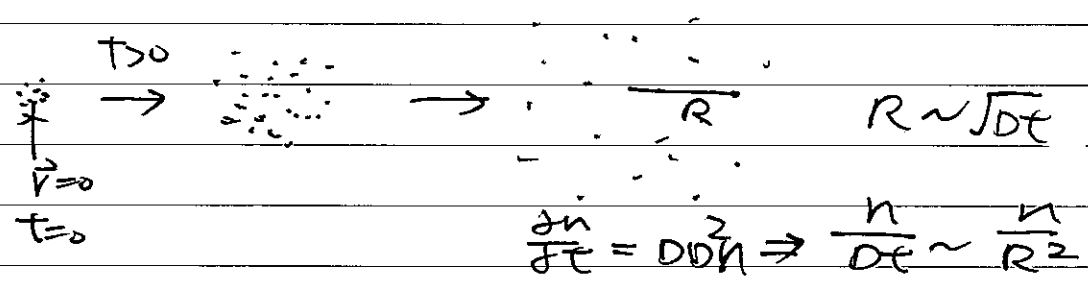
The diffusion equation originates the description for the spread of particles in the air or water.

$$\frac{dn}{dt} + \vec{\nabla} \cdot \vec{J} = 0 \quad n = \text{volume particle density}$$

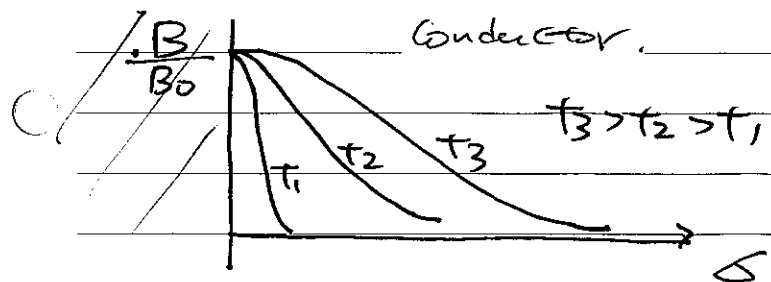
$$\vec{J} = -D \nabla n \quad D = \text{diffusion constant}$$

$$\frac{dn}{dt} = D \nabla^2 n$$

If at $t=0$, $n = f(\vec{r})$, n evolves as follows:



$\therefore \vec{E}$ & \vec{B} fields in the quasi-static approximation enter into a good conduction in the way diffusion



$\therefore D \propto \frac{1}{\sigma}$. \therefore if $\sigma \rightarrow \infty$, \vec{E} & \vec{B} can't penetrate a perfect conductor. σ impacts the diffusion of \vec{B} into a conductor.

In the quasi-static approximation, one has

the Kirchhoff's laws for circuits.

Here one sets $I = I(t)$, $\Sigma = \Sigma(t)$

$V = V(t)$

(i) Kirchhoff's current law:

In general, $\vec{\nabla} \times \vec{B} = \mu_0 (\vec{J} + \vec{J}_0)$

$$\therefore \vec{\nabla} \cdot (\vec{J} + \vec{J}_0) = 0$$

$\therefore \vec{J}$ itself would not conserve.

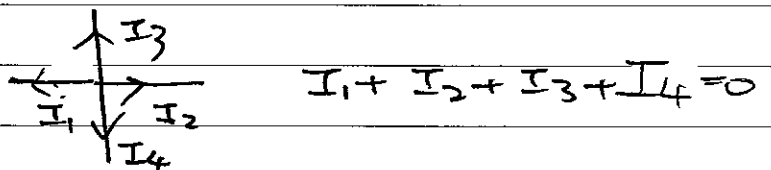
However, in quasi-static limit, \vec{J}_0 can be neglected (i.e. there is no ^{appreciable} charge

accumulation, $\vec{\nabla} \cdot \vec{J}_0 = \frac{d\rho}{dt}$)

$\therefore \vec{J}$ is conserved, i.e. $\vec{\nabla} \cdot \vec{J} = 0$

$\therefore \sum_i I_i = 0$ for each node

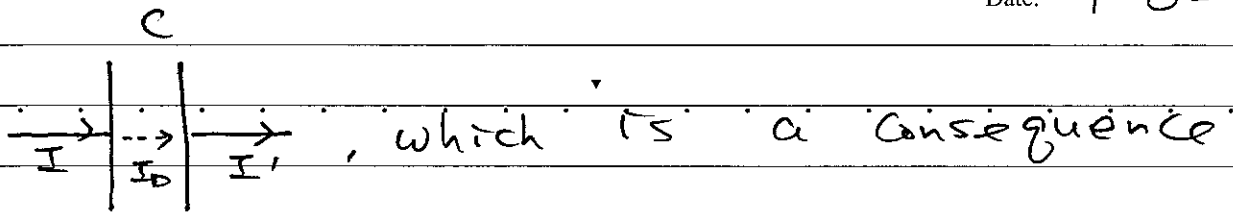
In circuit.



Note that the Kirchhoff's law does not

imply the conservation of current when the circuit breaks such as the current enter

= the current leave the capacitor.



of conservation of $\vec{J} + \vec{J}_0$: $I = I_0 \therefore I = I_1$
 $I_0 = I_1$

(ii) Kirchhoff's voltage law

In quasi-electrostatic approximation, $\vec{\nabla} \times \vec{E} = 0$

$$\therefore \oint \vec{E} \cdot d\vec{e} = 0 \quad \dots \text{iii)}$$

i.e. $\oint \vec{f}_E \cdot d\vec{e} = 0$ $\vec{f}_E =$ force per unit charge due to \vec{E} .

In applying the above approximation to AC

Circuits, one needs to introduce emf force

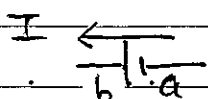
\vec{f}_s per unit charge either due to batteries

or induced emf.

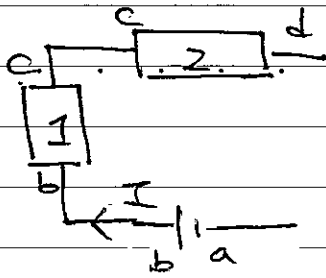
Inside emf devices, $\sigma = \infty$, $\therefore f = \frac{J}{\sigma} = 0$

$$\therefore \vec{f}_s + \vec{f}_E = 0 \quad (\text{for inductors, } \vec{f}_E = \vec{E} = -\frac{d\vec{A}}{f_E} = -\vec{f}_s \therefore \vec{f}_s = \frac{d\vec{A}}{f_E})$$

$$\therefore \int_a^b \vec{f}_E \cdot d\vec{e} = - \int_a^b \vec{f}_s \cdot d\vec{e} = \int_a^b \vec{E} \cdot d\vec{e} = V_a - V_b$$



$$\therefore V_a - V_b = -\mathcal{E} \quad ; \quad V_b - V_a = \mathcal{E} > 0$$



Therefore, (ii) implies $V_a - V_b + V_b - V_c + V_c - V_d + \dots = 0$

$$\therefore -\sum \Delta U_{\epsilon} + \Delta U_1 + \Delta U_2 + \dots = 0$$

where except for emf ϵ , $\Delta U_i = \bar{V}_i =$ voltage

drop across i th circuit element ($V_{\text{enter}} - V_{\text{leave}}$)

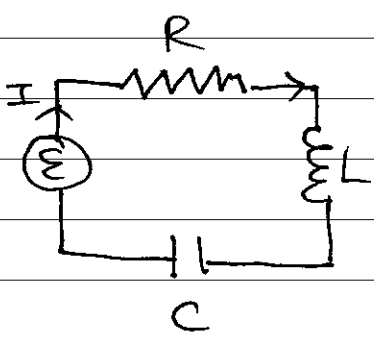
$$\epsilon = V_{\text{leave}} - V_{\text{enter}}$$

\therefore One has the Kirchhoff's voltage law

$$\sum_i \bar{V}_i = 0$$

$\bar{V}_i =$ voltage drop across i th circuit element.

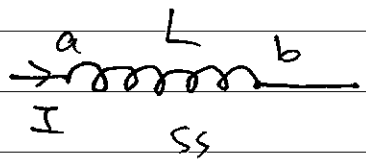
Example: RLC circuit.



$$-\epsilon + IR + L \frac{dI}{dt} + \frac{Q}{C} = 0$$

voltage drop across R
drop across C

$$\epsilon_L = -L \frac{dI}{dt} = \text{emf of inductance} = V_b - V_a$$



$$\therefore \epsilon = L \frac{dI}{dt} + IR + \frac{Q}{C}$$